

# EE1C1 “Linear Circuits A”

Week 1.8

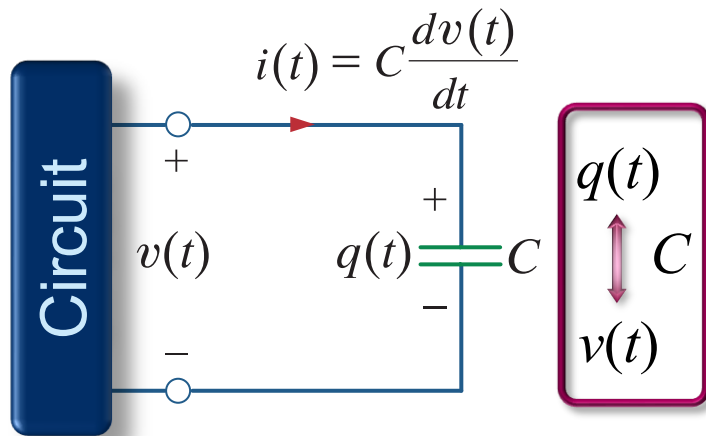
Francesco Fioranelli / Ioan E. Lager

# Today

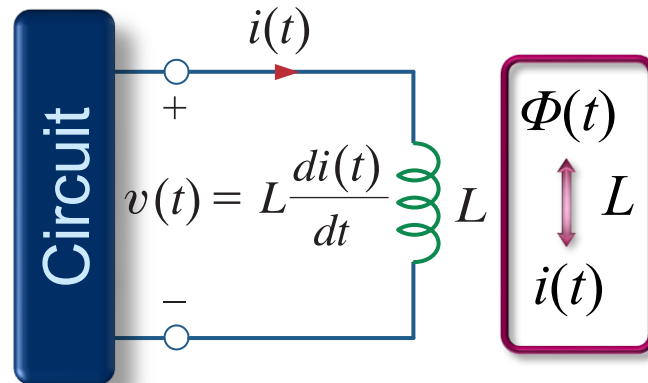
- Recap: Capacitance & capacitor; inductance & inductor;  $C$ – $C$  and  $L$ – $L$  interconnections
- New topics:
  - First-order transient circuits:  $RL$  and  $RC$
  - Step response
  - Exam exercise example
- Summary and conclusions
- Next tasks
- Your opinion counts!

# Recap of week 1.7

## Capacitance / capacitors



## Inductance / inductors








# Transient circuits

First-order circuits –  $RC$  and  $RL$

# Transient analysis

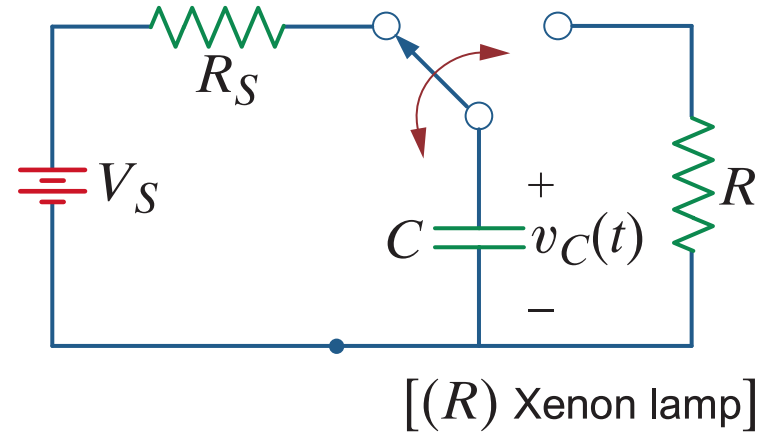
- Thus far we only examined the ‘steady state’
  - currents and voltages are constant
- Transient analysis amounts to examining circuits when a ‘disturbance’ is induced
- It is carried out in time-domain

# Transient analysis

- Capacitances and inductances can store and release energy
  - abrupt changes cannot have an instantaneous effect
  - the process depends on the rest of the circuit
- Time constant:
  - indicates how fast disturbances propagate through a circuit
  - light switch   thermostat 
- First-order circuits:
  - a combination of one resistance and one capacitance   $RC$
  - a combination of one resistance and one inductance   $RL$

# Transient analysis

- Camera with a flash



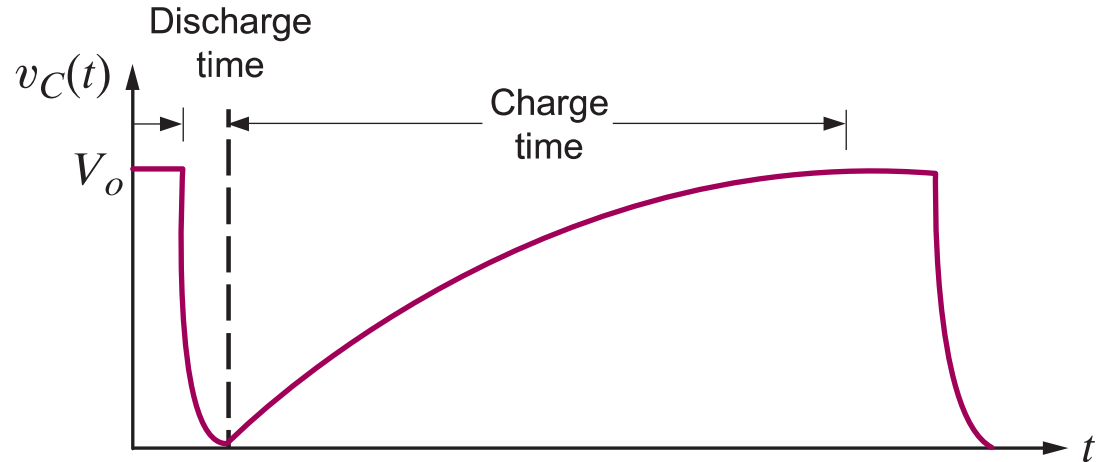
- Process:

- charging

- flash!



- charging



# Transient analysis

- The discharge part:
  - can be construed as a capacitance that releases its energy in a resistance  $R$
- KCL:

$$C \frac{dv_C(t)}{dt} + \frac{v_C(t)}{R} = 0$$



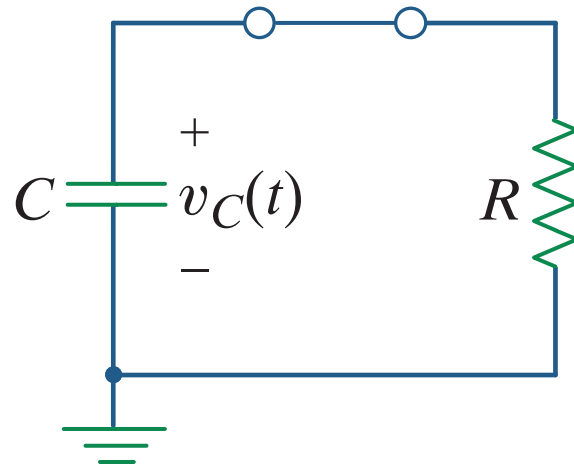
$$\frac{dv_C(t)}{dt} + \frac{1}{RC} v_C(t) = 0$$

- The solution:

$$v_C(t) = V_0 e^{-t/RC}$$



Differential equations:  
a math-in-a-nutshell moment





# Differential equations

- Analysing such circuits amounts to solving a differential equation

$$\frac{dv_C(t)}{dt} + \frac{1}{RC} v_C(t) = 0$$

- General form: 
$$\frac{dx(t)}{dt} + ax(t) = f(t)$$

- Task:** find  $x(t)$

# Differential equations

$$\frac{dx(t)}{dt} + ax(t) = f(t)$$

Linear, differential equations with constant coefficients  $\Rightarrow$  it holds:

- When one knows a (particular) solution for the **general form**

$$x(t) = x_p(t) \quad p \rightarrow \text{"particular"}$$

$$\frac{dx_p(t)}{dt} + ax_p(t) = f(t)$$

- **AND** the solution of the **homogeneous form**

$$x(t) = x_h(t) \quad h \rightarrow \text{"homogeneous"}$$

$$\frac{dx_h(t)}{dt} + ax_h(t) = 0$$

- **THEN** the complete solution of the differential equation is:

$$x(t) = x_p(t) + x_h(t)$$

# Differential equations

- For the time being, we only consider constant right-hand side terms of the form  $f(t) = A$

- General form:

$$\frac{dx_p(t)}{dt} + ax_p(t) = A$$

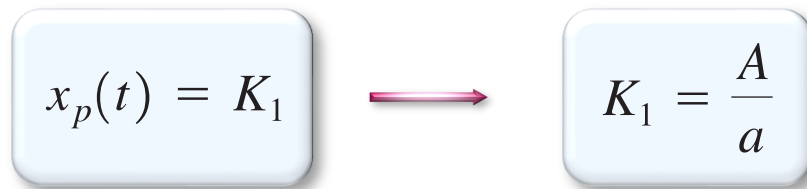
- Homogeneous form:

$$\frac{dx_h(t)}{dt} + ax_h(t) = 0$$

# Differential equations

- Since the right-hand side term is constant, we may assume that the particular solution  $x_p(t)$  is also constant

$$\frac{dx_p(t)}{dt} + ax_p(t) = A$$



The diagram illustrates the process of finding a particular solution. It consists of two light blue rounded rectangular boxes connected by a pink arrow pointing from left to right. The left box contains the equation  $x_p(t) = K_1$ . The right box contains the equation  $K_1 = \frac{A}{a}$ .

# Differential equations

- From the homogeneous equation it follows that:

$$\frac{dx_h(t)}{dt} + ax_h(t) = 0 \quad \longrightarrow \quad \frac{dx_h(t)/dt}{x_h(t)} = -a \quad \longrightarrow \quad \frac{d}{dt} [\ln x_h(t)] = -a$$

$c$  collects constants from both sides

- Consequently:

$$\ln x_h(t) = -at + c$$

$$x_h(t) = K_2 e^{-at}$$

$$K_2 = e^c$$

# Differential equations

- Combining the two solutions:  $x_p(t) = K_1 = \frac{A}{a}$  &  $x_h(t) = K_2 e^{-at}$



$$x(t) = x_p(t) + x_h(t) = \frac{A}{a} + K_2 e^{-at}$$

- $K_2$  can be determined if we knew  $x(t)$  at a given moment
- Problem solved!



# Differential equations

- In general, we can express the solution of our first-order differential equation as

$$x(t) = K_1 + K_2 e^{-t/\tau}$$

- $K_1$  is the “steady-state” or the stationary solution for  $t \rightarrow \infty$
- $\tau$  is the circuit’s **time constant**  $\longrightarrow$  it measures how fast the second term vanishes

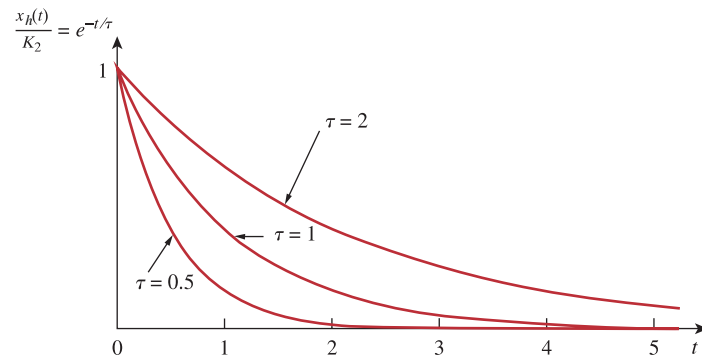
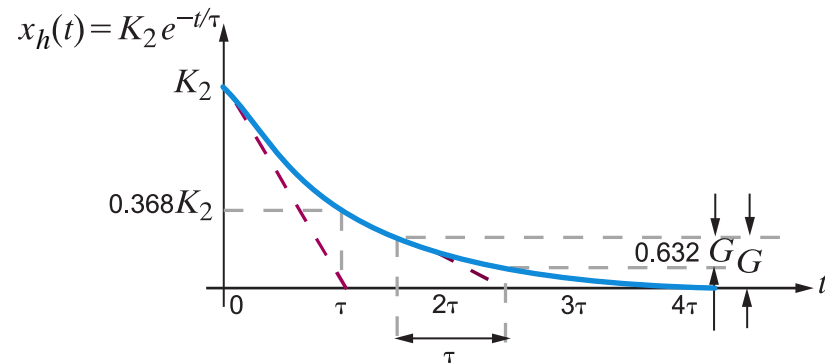
# Differential equations: time constant

- After each  $\tau$ , the quantity  $x_h(t)$  drops by  $e$  (Euler's number)

Values of  $x_h(t)/K_2 = e^{-t/\tau}$

$t$	$x_h(t)/K_2$
$\tau$	0.36788
$2\tau$	0.13534
$3\tau$	0.04979
$4\tau$	0.01832
$5\tau$	0.00674

- The smaller the time constant  $\tau$ , the more abrupt the transition





# Method 1: Differential equations

## + examples

# Algorithm

- Write the KCL/KVL equations pertaining to
  - the voltage over the capacitance
  - the current through the inductance
- Fill in the generic solution:

$$x(t) = K_1 + K_2 e^{-t/\tau}$$

- Identify the coefficients and make use of the initial values

# Example 1: loading of a capacitance

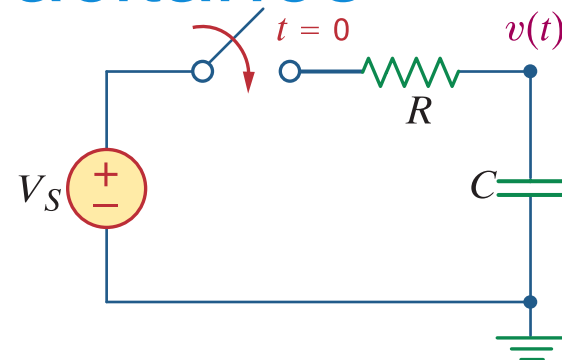
- KCL:  $C \frac{dv(t)}{dt} + \frac{v(t) - V_S}{R} = 0$

$$\frac{dv(t)}{dt} + \frac{v(t)}{RC} = \frac{V_S}{RC}$$

- Fill in the generic solution:

$$x(t) = K_1 + K_2 e^{-t/\tau}$$

$$-\frac{K_2}{\tau} e^{-t/\tau} + \frac{K_1}{RC} + \frac{K_2}{RC} e^{-t/\tau} = \frac{V_S}{RC}$$



# Example 1: loading of a capacitance

$$-\frac{K_2}{\tau}e^{-t/\tau} + \frac{K_1}{RC} + \frac{K_2}{RC}e^{-t/\tau} = \frac{V_s}{RC}$$

- The constant terms and the coefficients in the exponentials are:

$$K_1 = V_s$$

$$\text{and } \tau = RC$$

- Thus:  $v(t) = V_s + K_2 e^{-\frac{t}{RC}}$

- $K_2$  follows by observing the initial condition: the voltage over the capacitance is zero at  $t = 0 \rightarrow K_2 = -V_s$

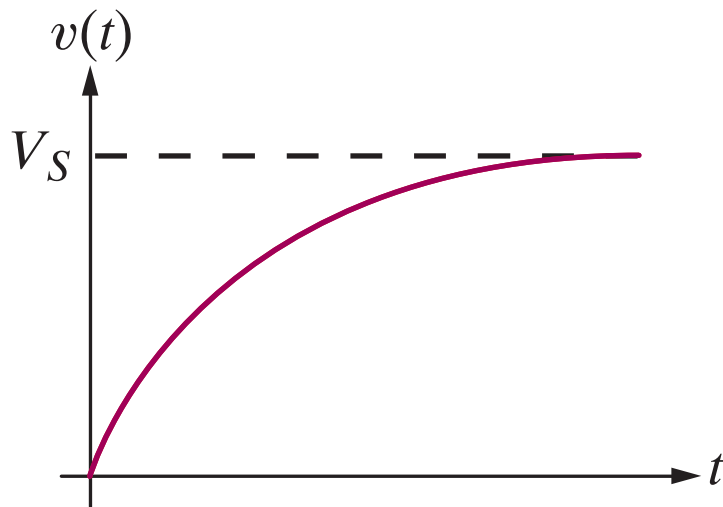
- We can now conclude that:

$$v(t) = V_s - V_s e^{-t/RC} = V_s (1 - e^{-t/RC})$$

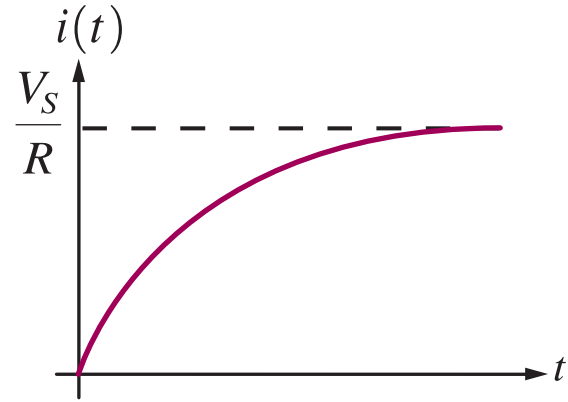
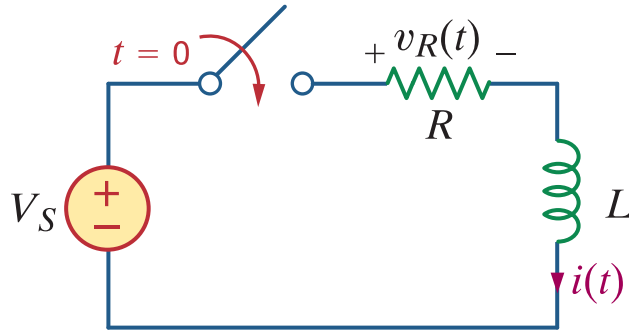
# Example 1: loading of a capacitance

- Plot of the solution

$$v(t) = V_S (1 - e^{-t/RC})$$



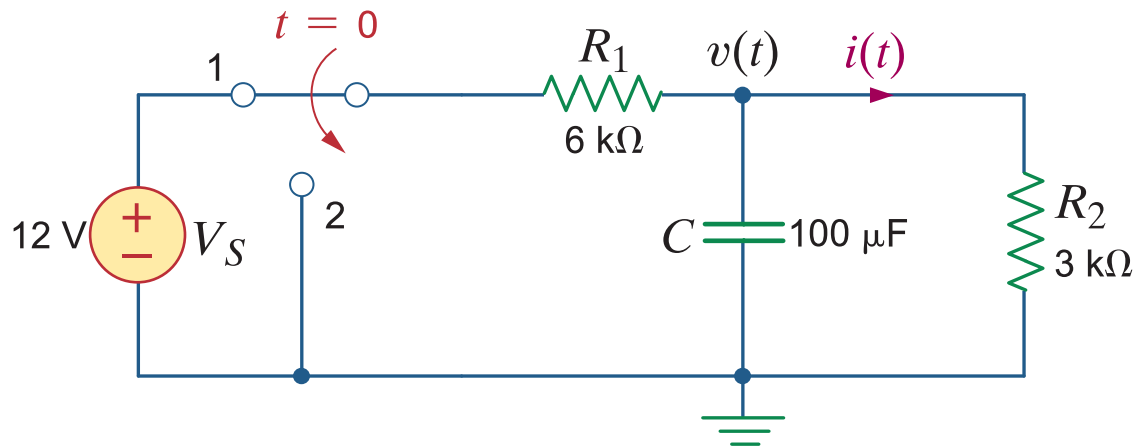
# *RL* circuit: practically the same



$$i(t) = \frac{V_S}{R} \left( 1 - e^{-\frac{R}{L}t} \right)$$

## Example 2: discharge of a capacitance

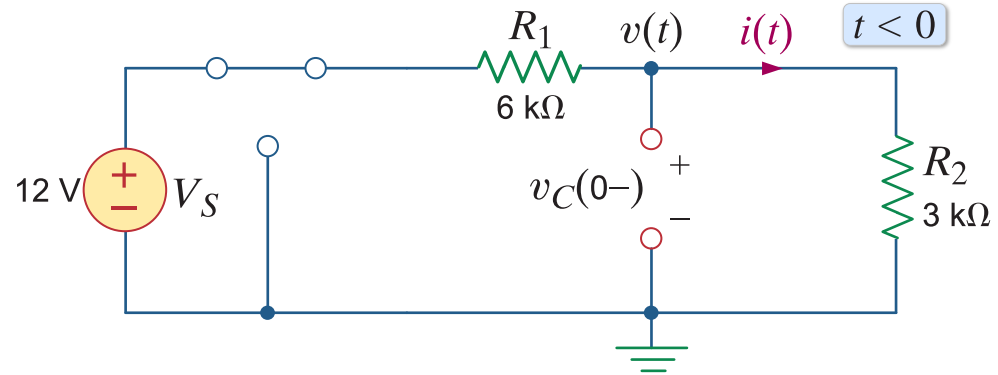
- The switch is taken to have been for very long in position 1



- Determine the current  $i(t)$  for  $t > 0$

# Example 2: discharge of a capacitance

- For  $t \uparrow 0$  we have:

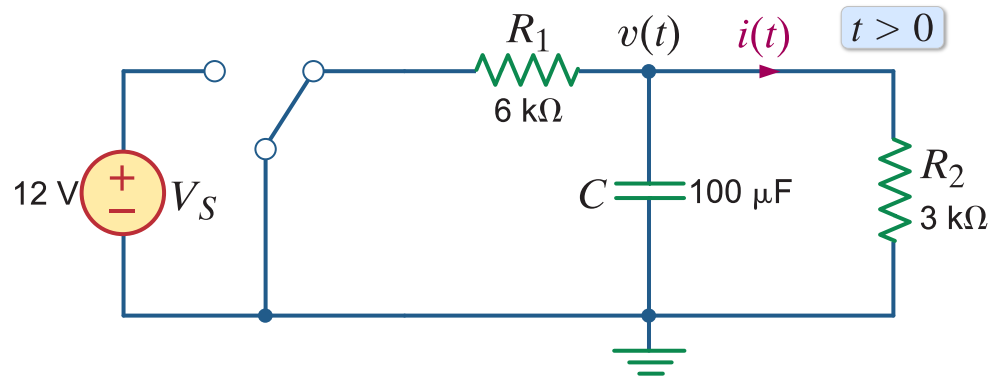


- The capacitance is fully charged  $\Rightarrow$  the current is zero
- The initial voltage is:  $v(0-) = 12 \left( \frac{3k}{6k + 3k} \right) = 4\text{ V}$



## Example 2: discharge of a capacitance

- For  $t > 0$  we have:



- KCL:  $\frac{v(t)}{R_1} + C \frac{dv(t)}{dt} + \frac{v(t)}{R_2} = 0$
- With the given values:  $\frac{dv(t)}{dt} + 5v(t) = 0$

- $R_1 = 6\text{ k}\Omega$
- $R_2 = 3\text{ k}\Omega$
- $C = 100\text{ }\mu\text{F}$

## Example 2: discharge of a capacitance

- This is a homogeneous equation:  $\frac{dv(t)}{dt} + 5v(t) = 0$
- The solution reads:  $v(t) = K_2 e^{-t/\tau}$
- Filling in the values yields:

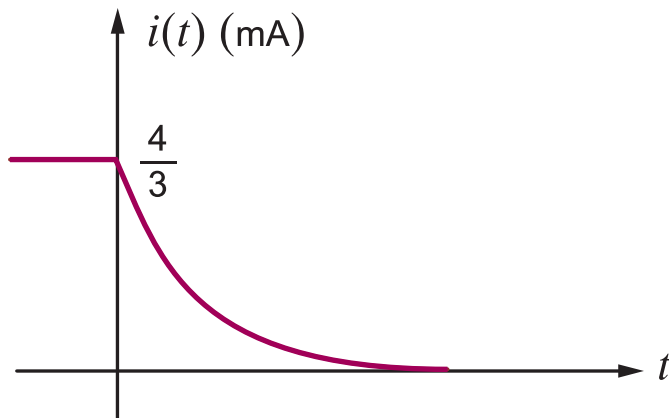
$$v(t) = K_2 e^{-t/0.2} \text{ V}$$

## Example 2: discharge of a capacitance

- Filling in the initial voltage  $v(0^-) = 4\text{V}$   $\longrightarrow$   $v(t) = 4e^{-t/0.2} \text{ V}$

- Thus, the current is:  $i(t) = \frac{v(t)}{R_2}$   $\longrightarrow$   $i(t) = \frac{4}{3}e^{-t/0.2} \text{ mA}$

- Plot:



# Method 2: Algorithmic

## + example

# Algorithm

- 1) Assume that the solution has the form:  $x(t) = K_1 + K_2 e^{-t/\tau}$
- 2) Assume that the circuit is in 'steady state' before the change occurs  $\Rightarrow$  express the current through  $L$  or the voltage across  $C$  **just before** the switch switches
- 3) Make use of the fact that the current or the voltage cannot be discontinuous:  $v_C(0+) = v_C(0-)$   
 $i_L(0+) = i_L(0-)$

# Algorithm

4) Redraw the circuit in the new 'steady state' that holds for  $t \rightarrow \infty \Rightarrow$  practically, this applies for  $t > 5\tau \Rightarrow$  solve the new circuit  $\Rightarrow x(t)|_{t>5\tau} = x(\infty)$

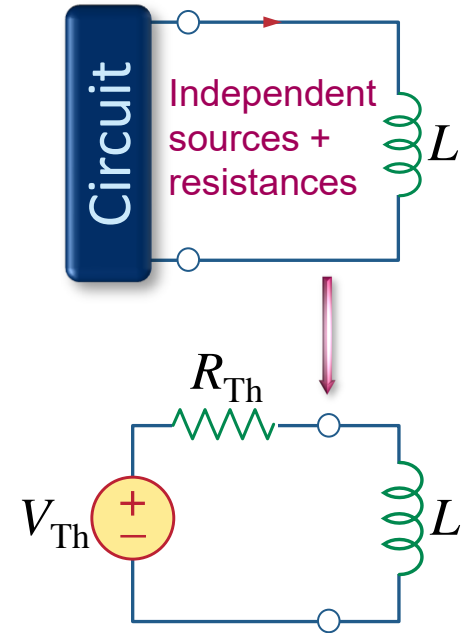
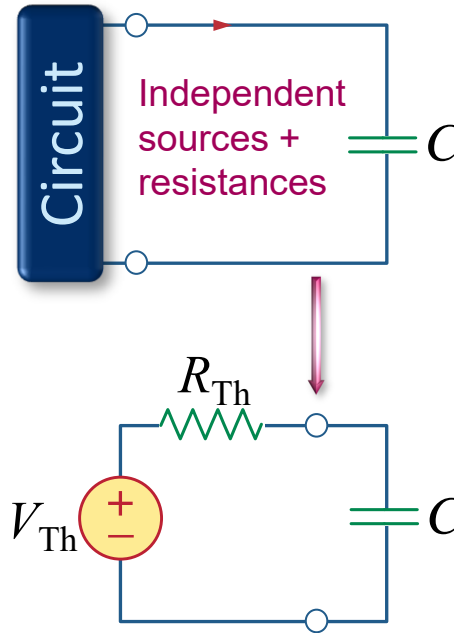
5) The time constant holds for all currents and voltages



It is then easier to reduce the circuit to a voltage source, a resistance and a reactive element (capacitance or inductance)  $\Rightarrow$  the Thévenin equivalent at the reactive element's terminals:  $\tau = R_{Th}C$  and  $\tau = L/R_{Th}$

# Algorithm

- Assume that “Circuit” contains only independent sources & resistances
- It can be replaced by its Thévenin equivalent
- The canonical  $RC$  or  $RL$  circuits!  $\longrightarrow$



standard solution

# Algorithm

- 6) The constants at step 1 are now determined via the results from the steps 3, 4 and 5  $x(0+) = K_1 + K_2$  and  $x(\infty) = K_1$

This entails  $K_1 = x(\infty)$  and  $K_2 = x(0+) - x(\infty)$

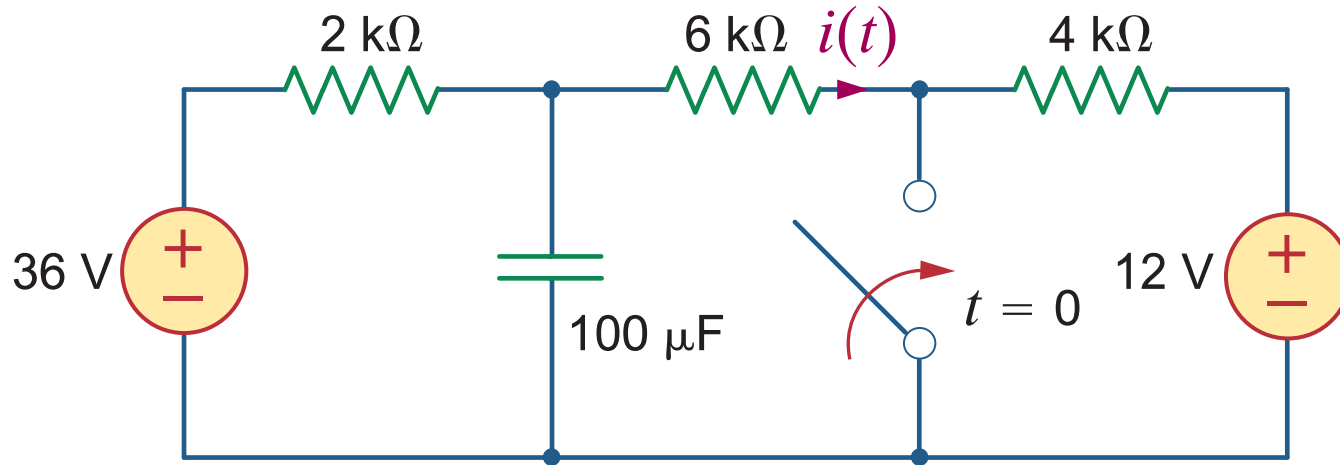
The final solution is:  $x(t) = x(\infty) + [x(0+) - x(\infty)]e^{-t/\tau}$

Initial – final values formula



# Example 3: complex circuit

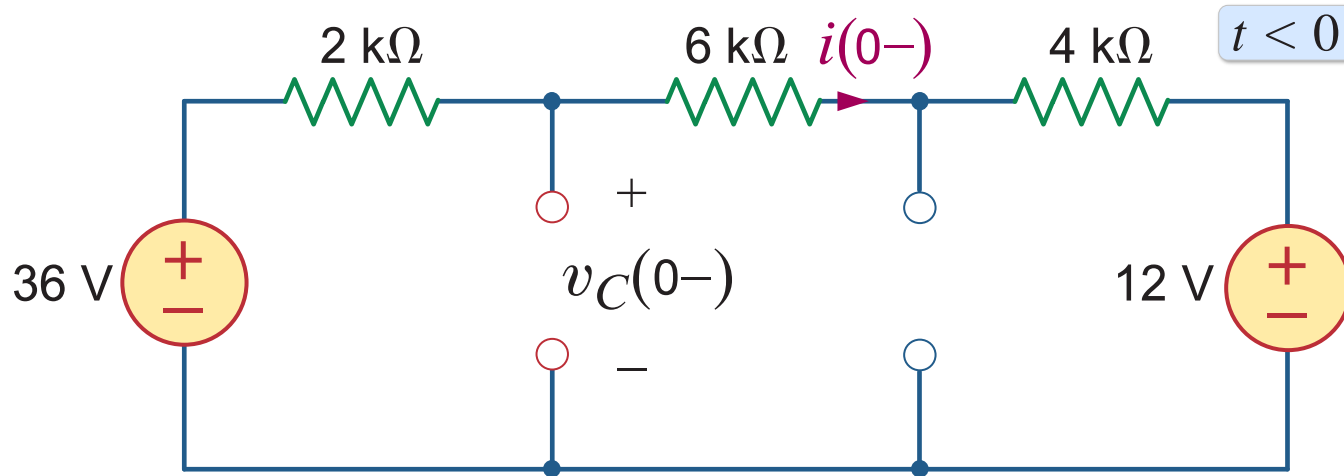
- Determine the current  $i(t)$  for  $t > 0$



- STEP 1:**  $i(t)$  is of the form:  $i(t) = K_1 + K_2 e^{-t/\tau}$

# Example 3: complex circuit

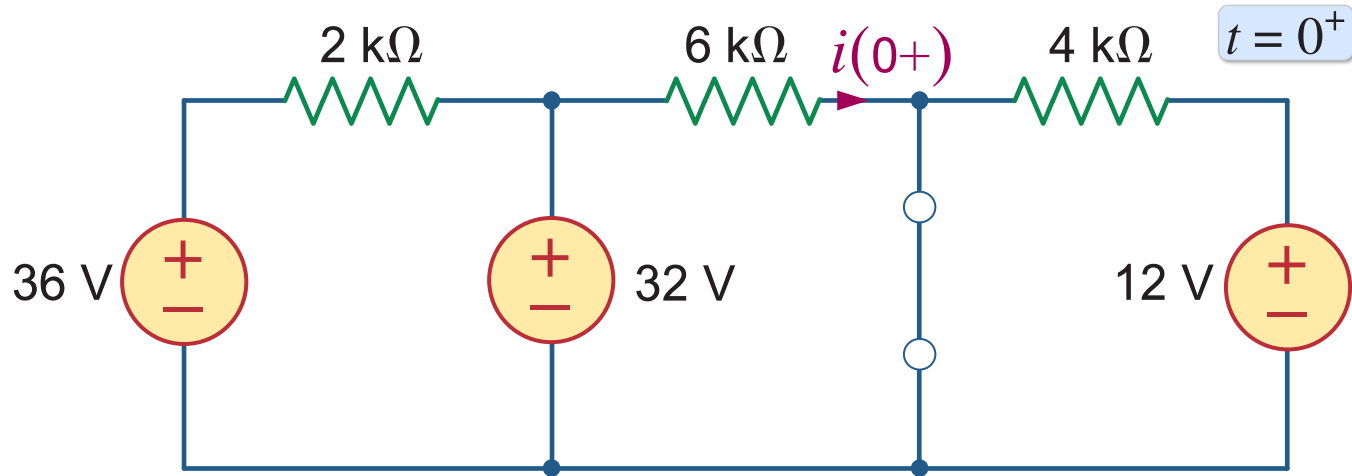
- STEP 2: Draw the circuit **just before** the switching moment



$$v_C(0-) = 36 - (2\text{k})(2\text{m}) = 32\text{ V}$$

# Example 3: complex circuit

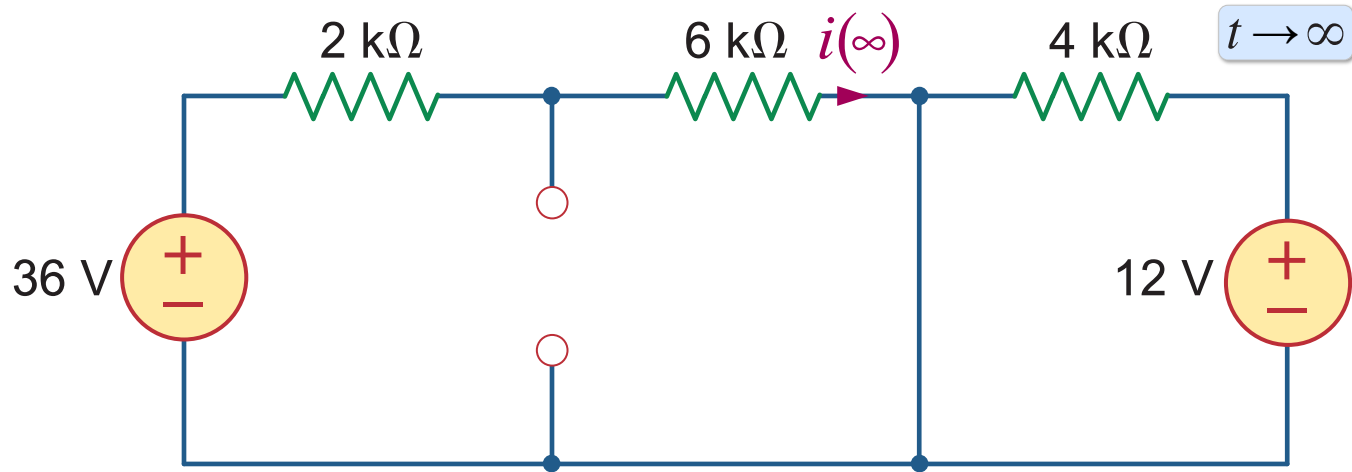
- STEP 2: Draw the circuit **just after** the switching moment



$$i(0^+) = \frac{32}{6k} = \frac{16}{3} \text{ mA}$$

# Example 3: complex circuit

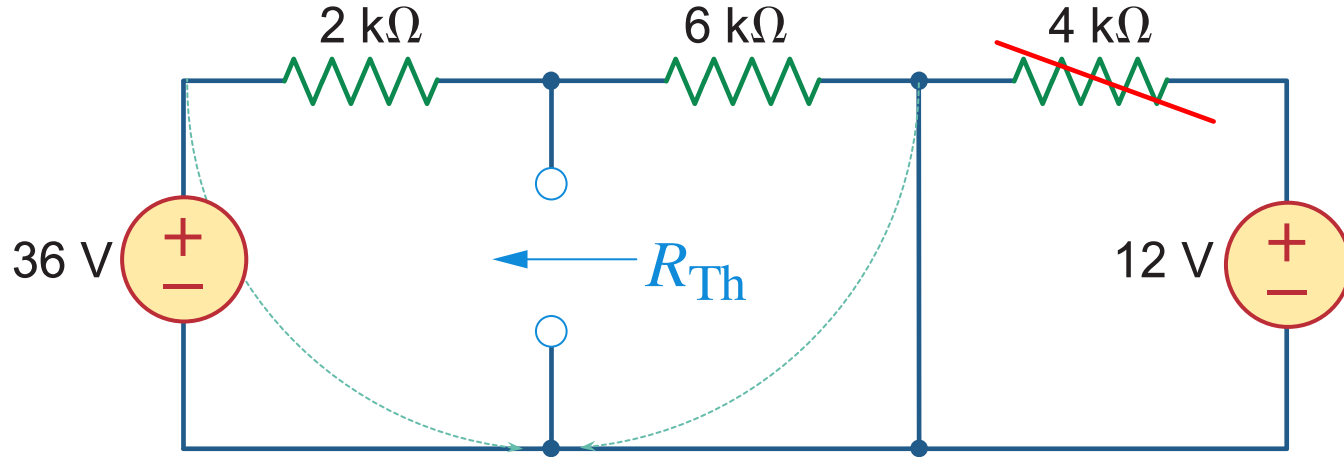
- **STEP 4:** The circuit in 'steady state', valid for  $t > 5\tau$



$$i(\infty) = \frac{36}{2k + 6k} = \frac{9}{2} \text{ mA}$$

# Example 3: complex circuit

- **STEP 5:** Determine  $R_{Th}$  at the capacitor's terminals



$$R_{Th} = \frac{(2k)(6k)}{2k + 6k} = \frac{3}{2} \text{ k}\Omega \quad \longrightarrow \quad \tau = R_{Th}C = \left(\frac{3}{2}\right)(10^3)(100)(10^{-6}) = 0.15 \text{ s}$$

## Example 3: complex circuit

- **STEP 6:** Make use of the steps 3, 4 and 5

$$K_1 = i(\infty) = \frac{9}{2} \text{ mA}$$

$$K_2 = i(0+) - i(\infty) = i(0+) - K_1 = \frac{16}{3} - \frac{9}{2} = \frac{5}{6} \text{ mA}$$

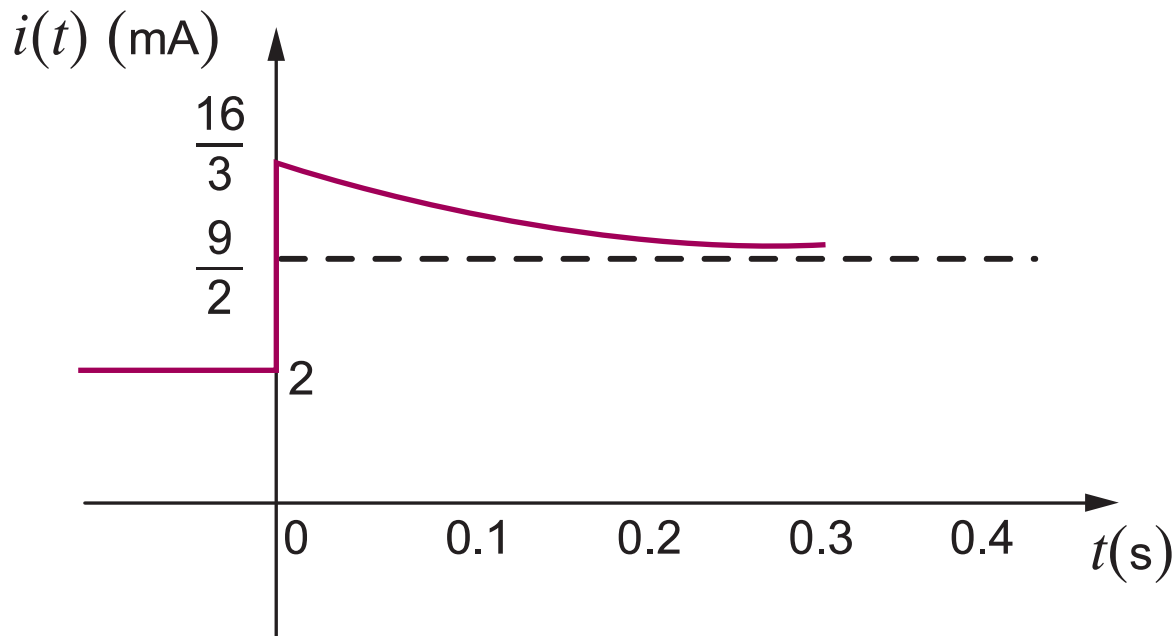
- Hence, it follows that:

$$i(t) = \frac{9}{2} + \frac{5}{6} e^{-t/0.15} \text{ mA}$$

# Example 3: complex circuit

- Plot of  $i(t)$

$$i(t) = \frac{9}{2} + \frac{5}{6} e^{-t/0.15} \text{ mA}$$



# Step response



# Step response

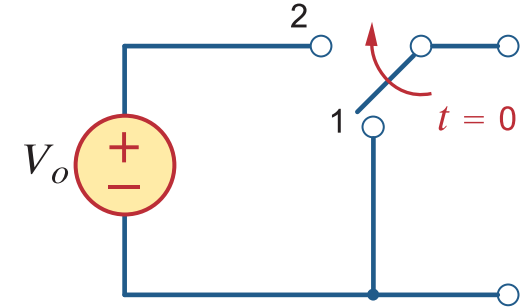
- Examples thus far: abruptly turning on or off a voltage or a current source
- Mathematical model: a unit step function

$$u(t) = \begin{cases} 0 & \text{for } t < 0 \\ 1 & \text{for } t > 0 \end{cases}$$



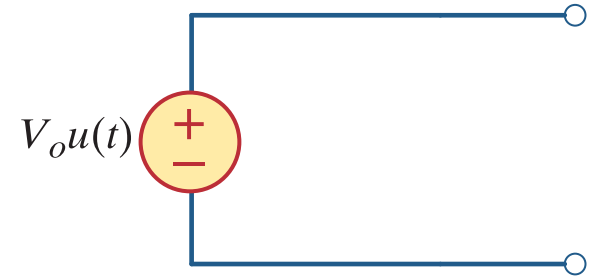
# Step response

- In a circuit: a voltage source + a switch



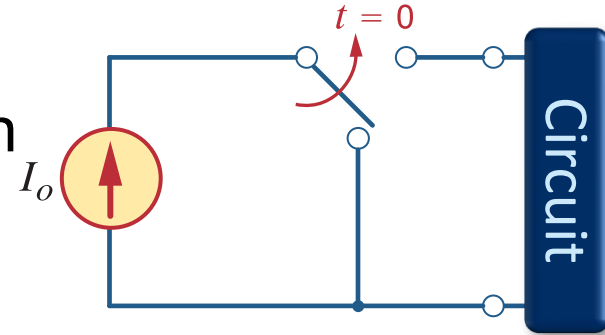
- = a voltage source  $V_o u(t)$ :

$$u(t) = \begin{cases} 0 & \text{for } t < 0 \\ 1 & \text{for } t > 0 \end{cases}$$



# Step response

- In a circuit: a current source + a switch



- = a current source  $I_o u(t)$ :

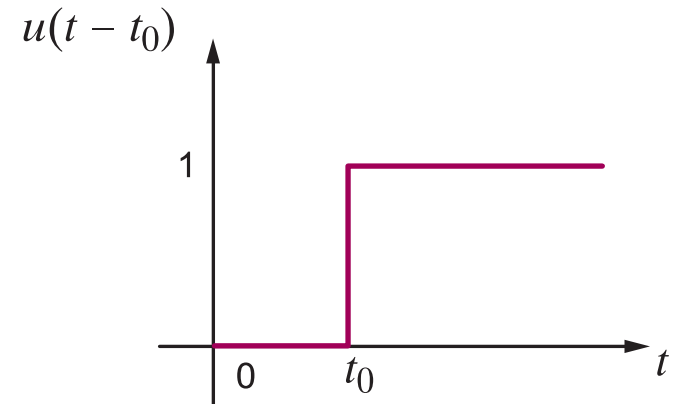
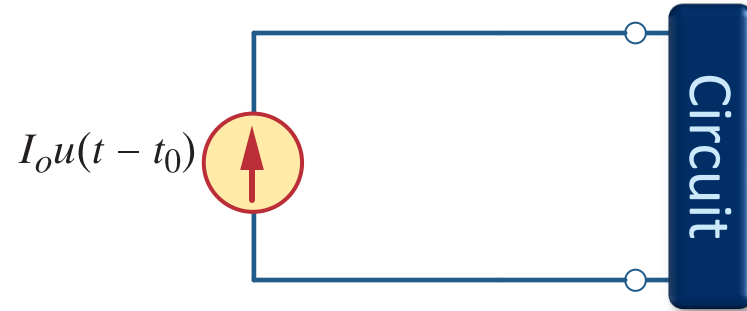
$$u(t) = \begin{cases} 0 & \text{for } t < 0 \\ 1 & \text{for } t > 0 \end{cases}$$



# Step response

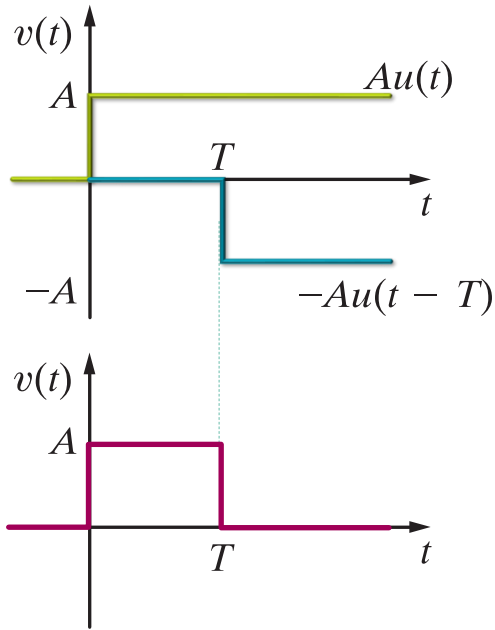
- A time shift can be added:
- The unit step function becomes:

$$u(t - t_0) = \begin{cases} 0 & \text{for } t < t_0 \\ 1 & \text{for } t > t_0 \end{cases}$$



# Step response

- With **two** unit step functions one can synthesise **a pulse**



$$v(t) = A[u(t) - u(t - T)]$$

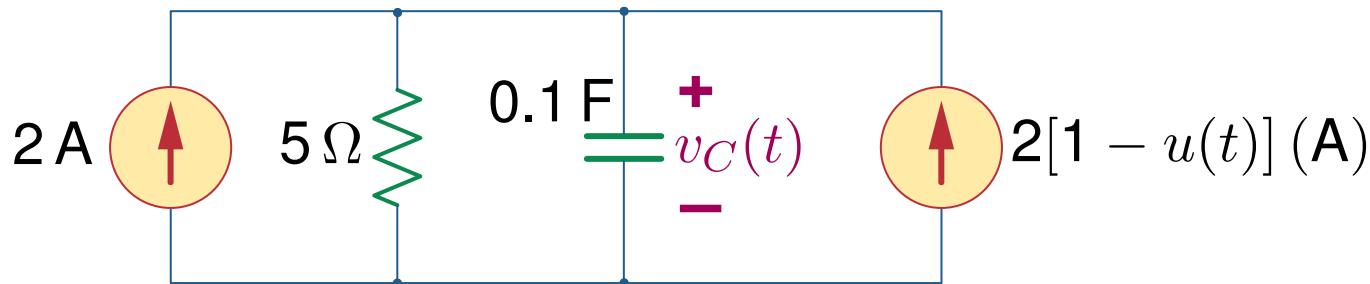
Or time-shifted by  $t_0$ :

$$v(t) = A\{u(t - t_0) - u[t - (t_0 + T)]\}$$

# Exam exercise example

# Exam(ple)

- Consider the following circuit:

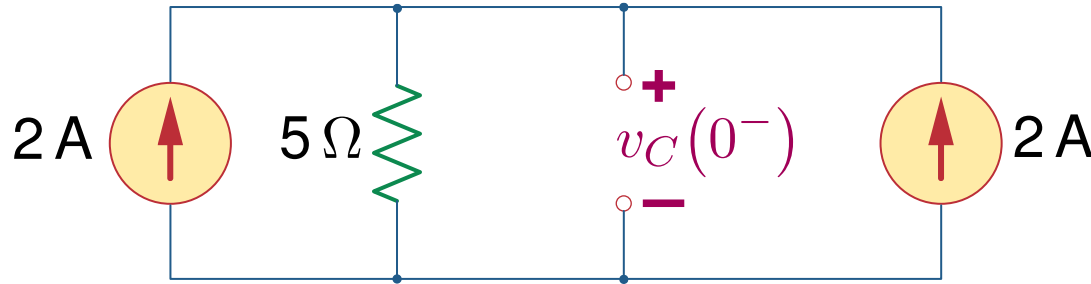


- Determine  $v_C(0+)$ .
- Determine  $v_C(t)$  for  $t > 0$  s.

# Exam(ple)

a) Determine  $v_C(0+)$ .

- Redraw the circuit for  $t < 0$ :



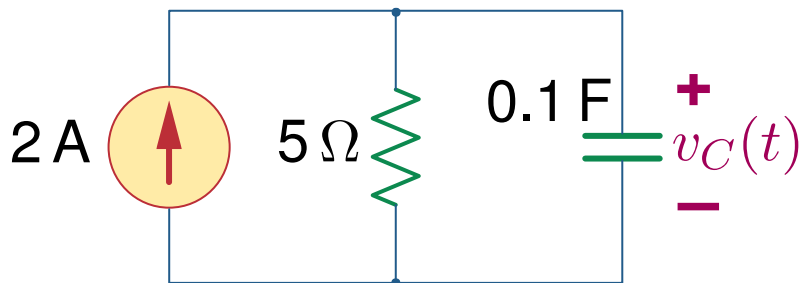
- Calculate the voltage  $v_C(0^-)$ :  $v_C(0^-) = (2 + 2)5 = 20\text{ (V)}$
- Apply the continuity conditions at  $t = 0$ :  
$$v_C(0^+) = v_C(0^-) = 20\text{ (V)}$$



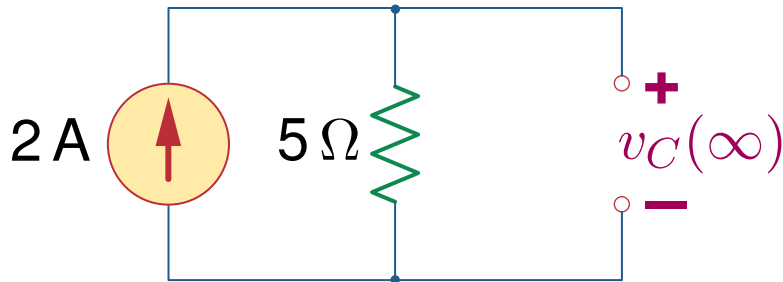
# Exam(ple)

b) Determine  $v_C(t)$  for  $t > 0$  s.

- Redraw the circuit for  $t > 0$ :



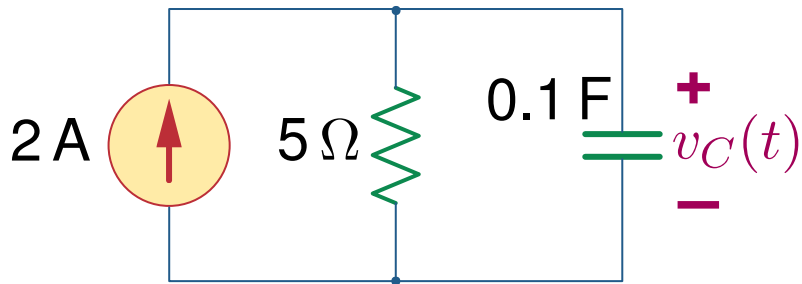
- Redraw the circuit for  $t \rightarrow \infty$ :



- Calculate  $v_C(\infty)$ :  $v_C(\infty) = 2 \cdot 5 = 10$  (V)

# Exam(ple)

b) Determine  $v_C(t)$  for  $t > 0$  s.



- Calculate the Thévenin resistance:  $R_{Th} = 5 \Omega$
- Calculate the time constant:  $\tau = R_{Th}C = 0.5$  (s)
- Assemble  $v_C(t)$  via the initial-final values formula:

$$\begin{aligned} v_C(t) &= v_C(\infty) + [v_C(0^+) - v_C(\infty)] \exp(-t/\tau) \\ &= 10 (1 + e^{-2t}) \text{ (V)} \end{aligned}$$

# Summary of the day

- First-order transient circuits:
  - always one capacitance or one inductance
  - can contain more resistances and/or sources
- The solution can be derived via:
  - differential equation techniques + identifying coefficients
  - an algorithm
- Step response:
  - step function  $u(t)$
  - pulses can be assembled by combining 2 step functions

# Next tasks

- Please do the SGH6
- Seminars of Tuesday and Friday
- Register for the end-of-term exam!
- Next week: second-order circuits

Thank you!

# Your opinion counts!

- The course management and we are highly interested in your feedback on the course → let us know what you think about the methods of education, assessment, organisation, etc.
- We are also interested in your opinion on the Circuits Course Labs
- Please go to: <https://evasys-survey.tudelft.nl/evasys/online.php?p=4EXUN>,

or scan the QR-code



Many thanks in advance for your feedback!

The results will be published on Brightspace

Questions: [QualityAssurance-EEMCS@tudelft.nl](mailto:QualityAssurance-EEMCS@tudelft.nl)